# **MEMS Toroidal Inductors for Integrated Power Electronics**

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## Introduction

- Increasing power density in power electronics circuits requires higher frequencies. As switching frequencies rise, the size of the magnetics fall, and new fabrication strategies such as MEMS fabrication become possible.
- Our goal is to integrate the magnetics into the power electronics circuits to achieve yet higher power densities.
- Focus on toroidal inductors since they produce minimal external fields, external losses and electromagnetic interference problems.
- Magnetics can be on insulator or siliconembedded.

# **Toroidal Magnetics**

**On-Insulator** Inductors





In-Silicon Inductors







| <ul> <li>Circuit and magnetics co-optimization requires equivalent circuit models for the magnetics.</li> <li>Different models are required for magnetics built on top of an insulating substrate as opposed to embedded in the silicon substrate.</li> </ul>  | Field Lines   |
|--|---|
| Closed-Form Solutions for Losses   | Equivalent Circuit  |
| Analytical models are developed for electrically-driven and magnetically-driven losses, and parasitic capacitance.<br>$\hat{\mathbf{r}} \leftarrow \hat{\mathbf{r}}^{2}  P_{M2} = \sigma  \omega^{2}  \mu_{o}^{2}  N^{2}  I^{2}  D^{3}  \ln(R_{o}/R_{i})  /  (48  \pi)$  | $= \underbrace{\begin{array}{c} \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\$   |
| $P_{E3} = \sum_{m=1}^{\infty} \frac{2\pi D \sigma V^2 \omega^2 \epsilon_1^2}{m (\Gamma_{\omega \epsilon}^2 + \Gamma_{\sigma}^2)}$ $P_{E3} = \sum_{m=1}^{\infty} \frac{2\pi D \sigma V^2 \omega^2 \epsilon_1^2}{m (\Gamma_{\omega \epsilon}^2 + \Gamma_{\sigma}^2)}$ $\Gamma_{\omega \epsilon} = \omega \left(\frac{R_o}{R_o + \Delta}\right)^m (\epsilon_I - \epsilon_{Si})$ $+ \omega \left(\frac{R_o + \Delta}{R_o}\right)^m (\epsilon_I + \epsilon_{Si})$ | $V^{2}/2 \left(P_{E1} + P_{E3} + P_{E4}\right)$ $V^{2}/2 \left(P_{E1} + P_{E3} + P_{E4}\right)$ $V^{2}/2 \left(P_{E1} + P_{E3} + P_{E4}\right)$ |

 $P_{\rm M1} = \sigma W \,\omega^2 \,L_{\rm P}^2 \,I^2 / (16 \,\pi) \,/ \,P_{\rm M3} = D \,\sigma \,\omega^2 \,L_{\rm p}^2 \,I^2 / (8 \,\pi)$  $\Gamma_{\sigma} = \sigma_{\rm Si} \left( \left( \frac{R_{\rm o} + \Delta}{R_{\rm o}} \right)^m - \left( \frac{R_{\rm o}}{R_{\rm o} + \Delta} \right)^m \right)$  $C = 2 \pi F \epsilon_{\rm I} \left( R_{\rm o} D + R_{\rm i} D + 0.5 (R_{\rm o}^2 - R_{\rm i}^2) \right) / (N \Delta)$ 

## **Measurement Vs. Models**



# Loss Distribution Silicon Loss- High Resistivity 3.5 Copper Loss - High Resistivity Silicon Loss - Standard Resistivity 2.5 **[Natts]** 2 1.5 Copper Loss - Standard Resistivity

# 0.5

## Conclusions

 $2(P_{M1} + P_{M2} + P_{M3})/I^2$ 

 $C_{eq} = C \left( N^2 - 1 \right) / (12 N)$ 

- The equivalent circuit model matches well the measured behavior of silicon-embedded and on-insulator toroidal inductor.
- Silicon-embedded toroidal inductors are viable for integrated power electronics.
- As long as the silicon is removed from the toroidal core of the inductor at high doping densities, the close proximity of silicon need not greatly degrade inductor quality factor.
- Winding loss dominates the quality factors modeled and measured here for frequencies up to 40 MHz, making winding optimization



Other experiments have demonstrated 690 nH for an on-insulator inductor of similar style.

A maximum quality factor of 20 is measured for an in-silicon inductor (high resistivity wafer).

Frequency [MHz]

90

References

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more important than silicon selection

For higher frequencies, quality factor is a strong function of silicon resistivity as expected, making its selection more important than winding design. At these frequencies, electrically-driven silicon losses appear to be dominant, making large resistivity an important fabrication objective.

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