### Mixed-Signal IC Controllers and Low-Volume SMPS Topologies

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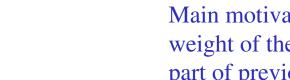
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### A Low Power (Portable) Application – Rogers Portable Radio





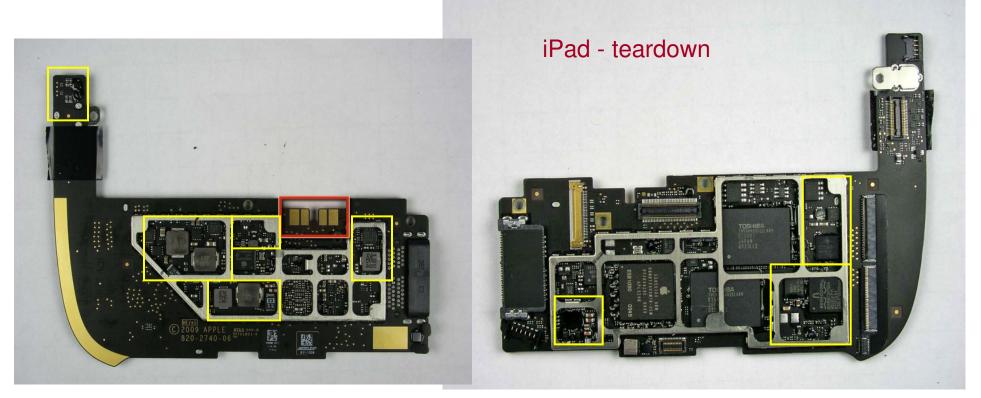




- Allowed a large number of households to have daily access to information
- Reduced power consumption by using more efficient electronic tubes

Main motivation: to reduce the volume and weight of the power supply, by far the largest part of previous radios

### Low Power SMPS in Portable/Consumer Electronics



http://www.ifixit.com/Teardown/iPad-FCC-Teardown/2197/1

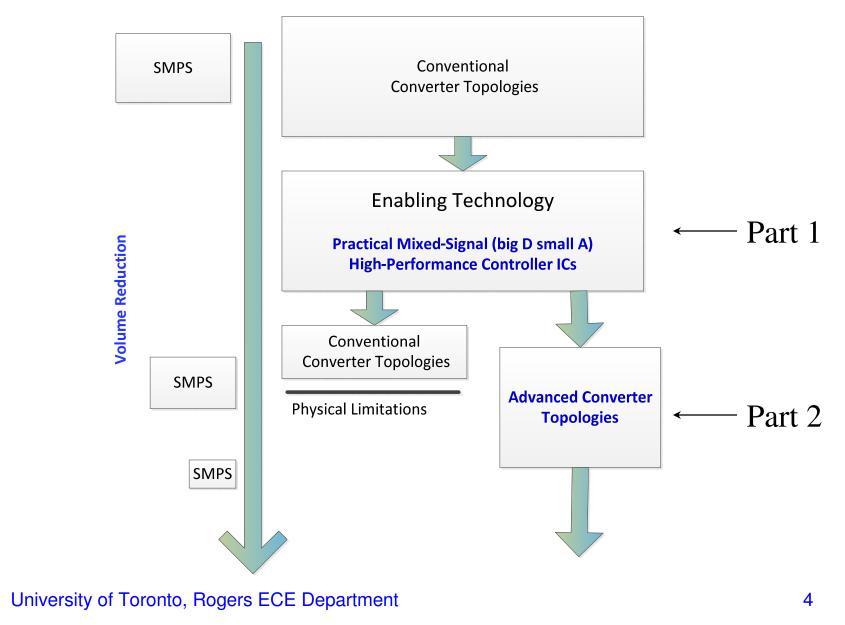


- Dc-dc SMPS occupy between 20% and 80% of the total volume in modern electronics devices, communication equipment, computers...
- Most of the volume occupied by passives and heat sinks

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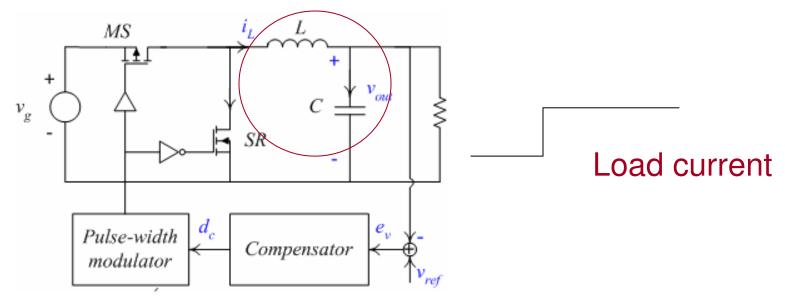
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### **Outline: Two Parts of Presentation**



### **Conventional Methods for Volume Minimization**

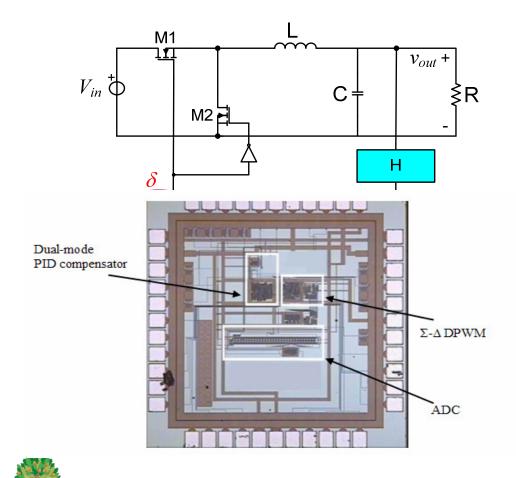
- Operation at high switching frequencies
- Efficiency optimization = flat and high efficiency curve
- Fast dynamic response of the controller + plug and play operation





All three of these goals can be acomplished

### **Ultra High-Frequency High-Resolution Digital Controller IC**



• The duty ratio of a core resolution (high-frequency) DPWM is varied over several switching cycles to achieve high effective resolution.

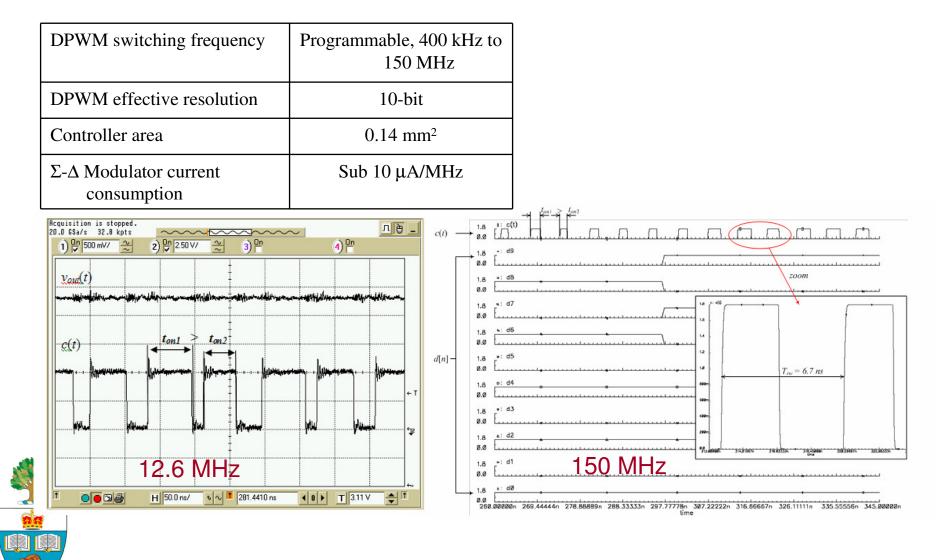
- The averaging is performed by the switching converter itself (LC filter)
- The output measured with a moving windowed ADC producing just few error signals
- Processing unit reduced since it also operates over a small error

#### range

[1] Z. Lukic, N. Rahman, A. Prodić, "Multibit  $\Sigma - \Delta$  PWM Digital Controller IC for DC–DC Converters Operating at Switching Frequencies Beyond 10 MHz," IEEEE Transactions on Power Electronics, Vol.22, Issue.5, October 2007, Pg. 1693-1707 (.pdf).

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### **Ultra High-Frequency Digital Controller IC Results**

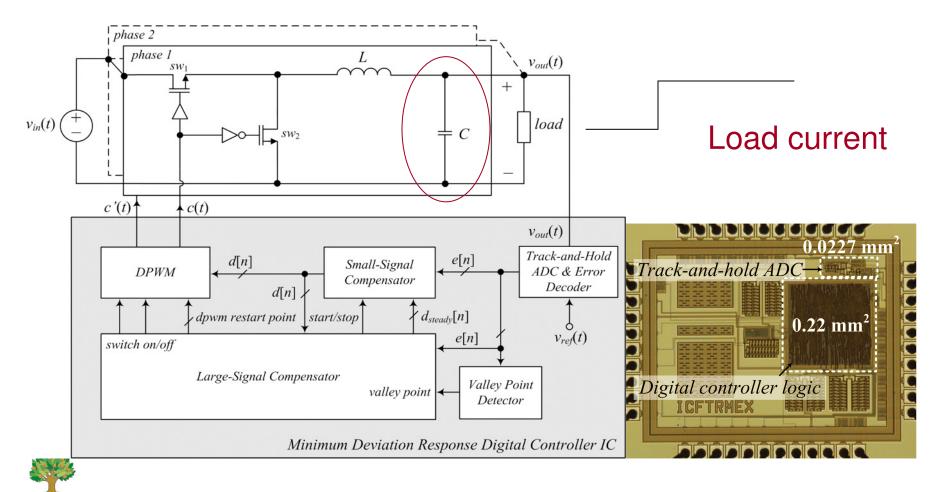


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### **Minimum Deviation Controller IC**<sup>[2]</sup>



[2] A. Radic, Z. Lukic. A. Prodic, and R. de Nie, "Minimum Deviation Digital Controller IC for DC-DC Switch-ModePower Supplies," IEEEE Transactions on Power Electronics, Vol.28, Issue.2, February 2013.

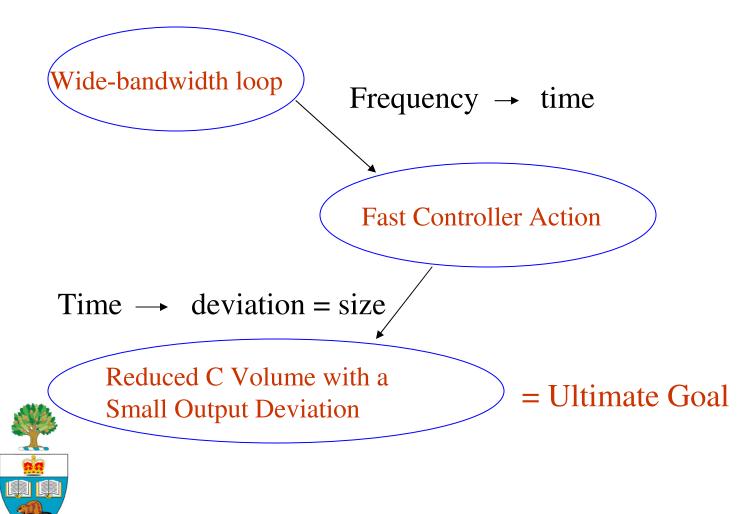
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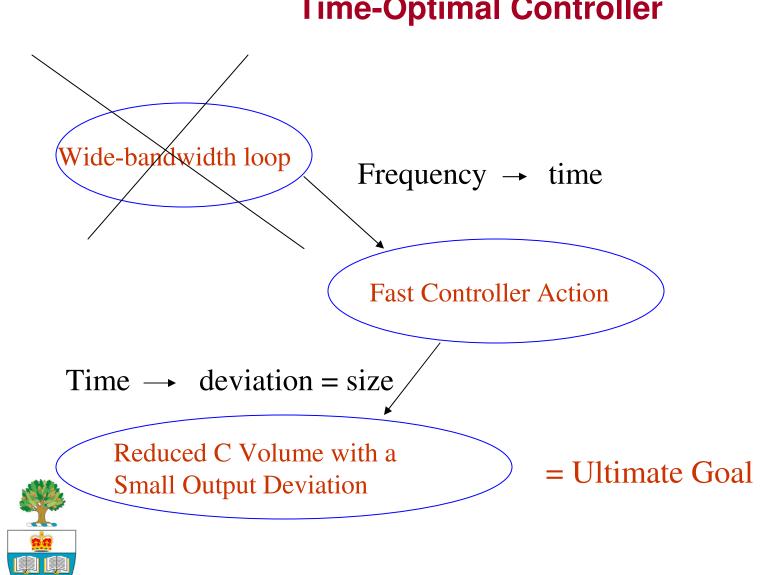
### **Conventional Controller Design**



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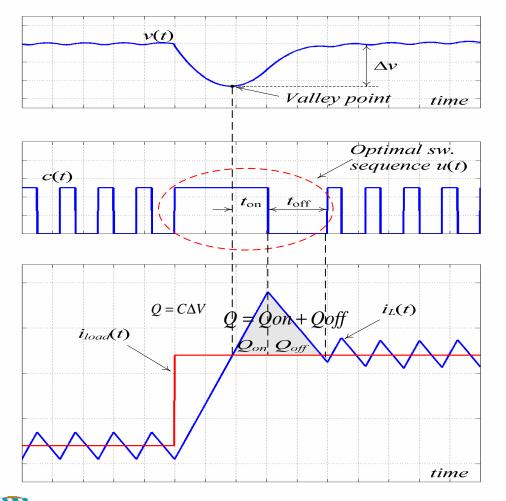


#### **Time-Optimal Controller**

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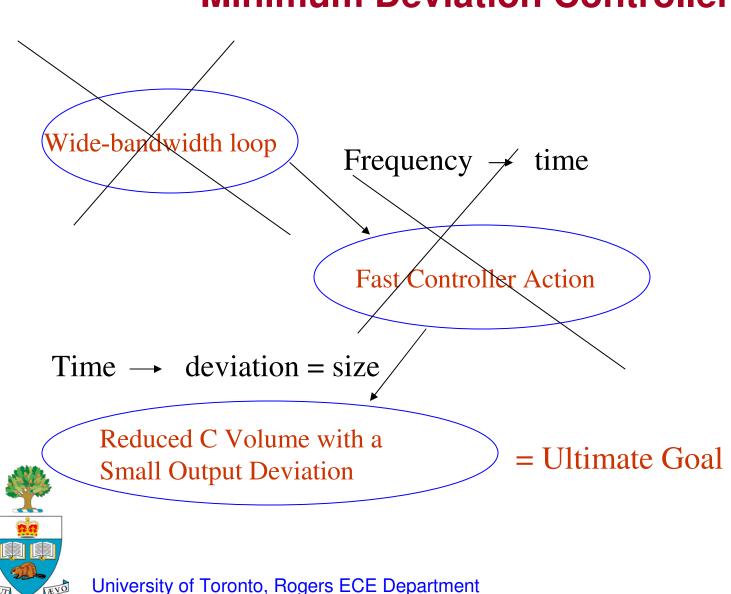
### **Time-Optimal Control**



Overly large peak inductor current, i.e. inductor might need to be overdesigned

- □ Fairly complex calculations
- Need to know LC values
- Very sensitive to time delays

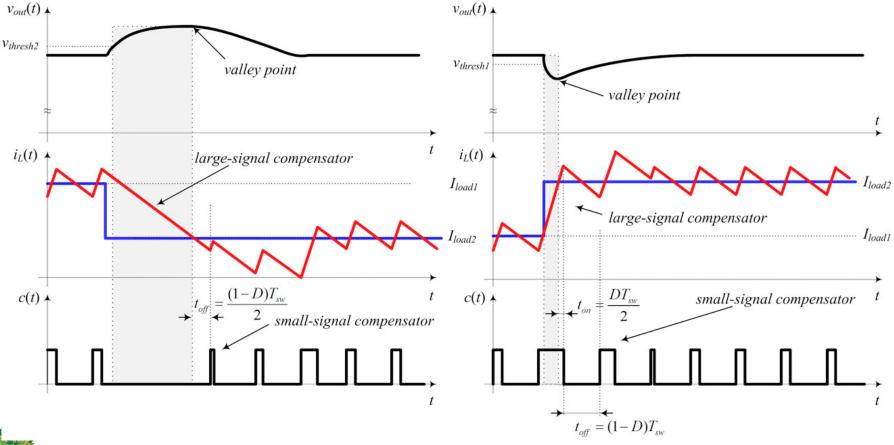




### **Minimum Deviation Controller**

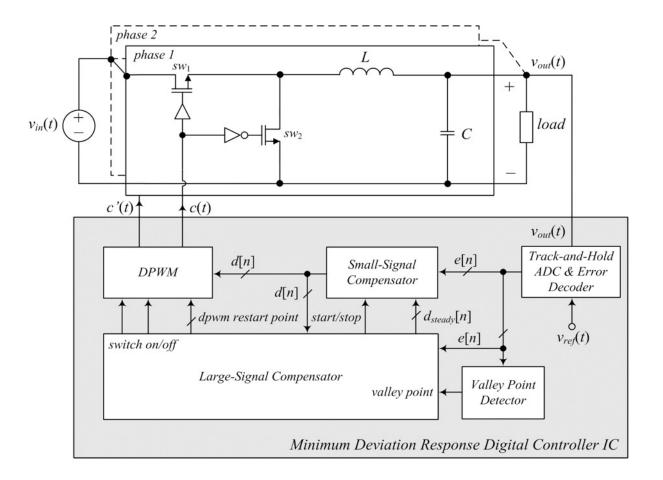
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### **Minimum (Optimum) Deviation Controller**



•No current overshoot, no need to know converter parameters, simple calculations

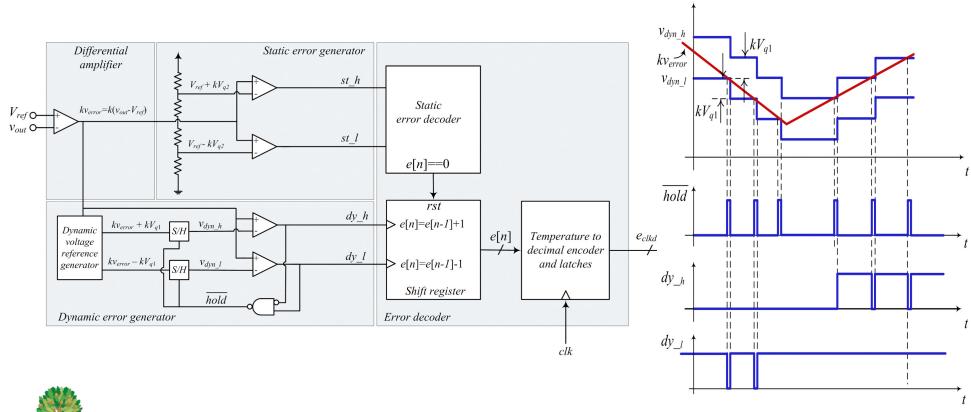
#### **Minimum (Optimum) Deviation Controller**





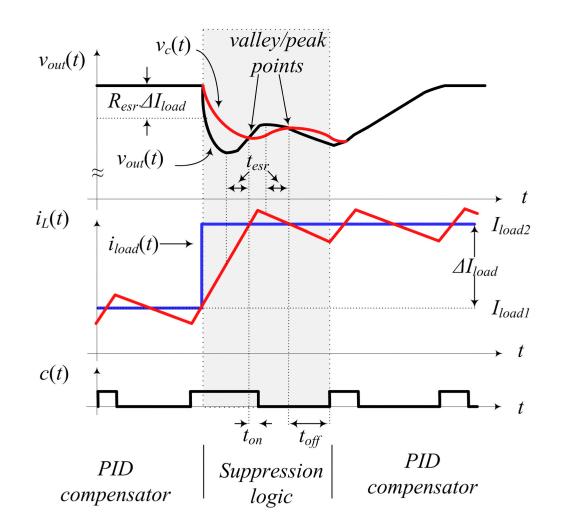
#### Only needs to remember *D* before transient

### Self-Calibrating SAR Track & Hold ADC



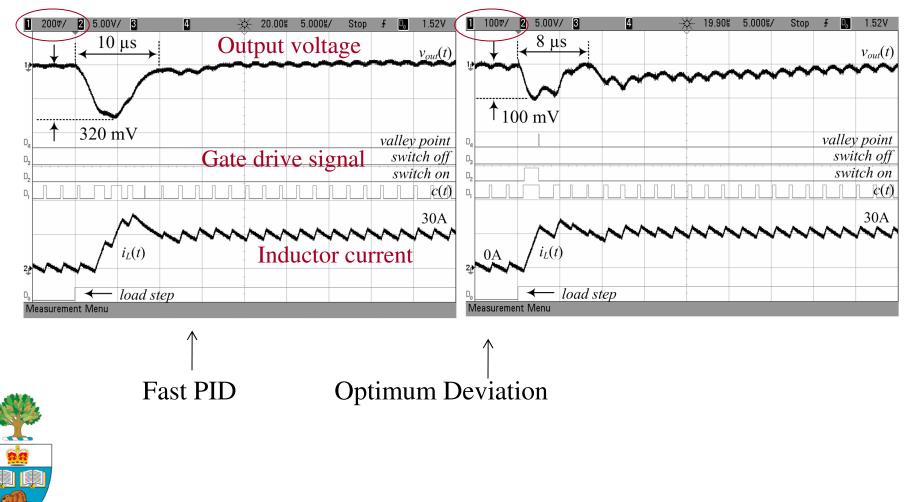


### **Dual Sampling Mechanism**





### **Practical Implementation (500 kHz VRM)**



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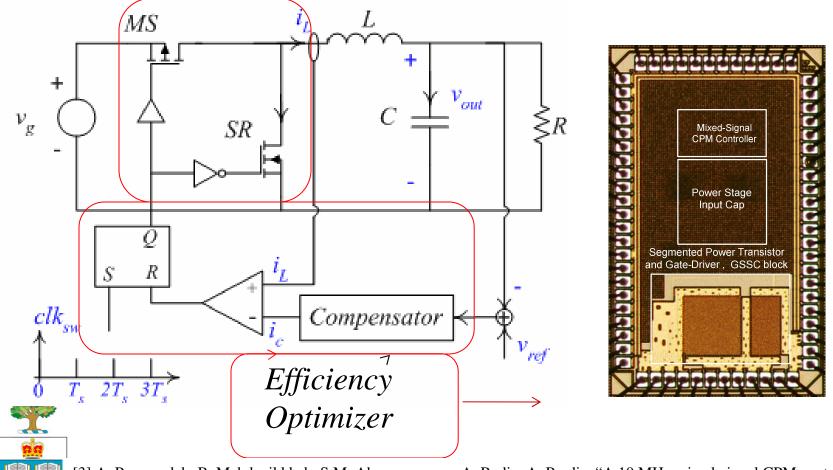
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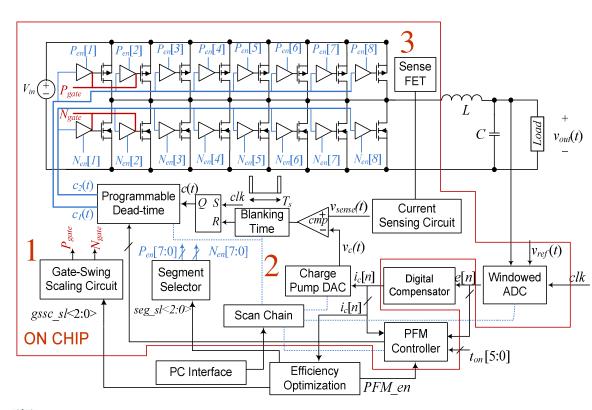
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### 10 MHz Mixed Signal CPM Power Module with Instantaneous Efficiency Optimization <sup>[3]</sup>



[3] A. Parayandeh, B. Mahdavikkhah, S.M. Ahsanuzzaman, A. Radic, A. Prodic, "A 10 MHz mixed-signal CPM controlled DC-DC converter IC with novel gate swing circuit and instantaneous efficiency optimization," in Proc. IEEE ECCE, 2011, Pg. 1229-1235

### **IC Block Diagram**

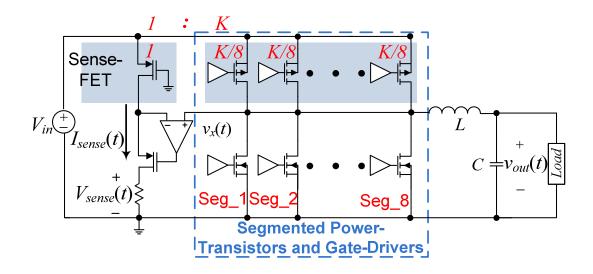


Incorporates:

- *1. Novel gate swing circuit*
- 2. Modified highresolution charge-pump based DAC
- 3. Optimized design of current sensing circuit (senseFET)



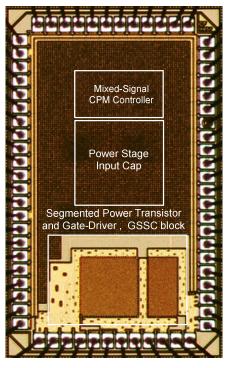
### Low-Power High-Frequency SensFET



*GBW reduced since the amplitude is always relatively large (but not the losses)* 



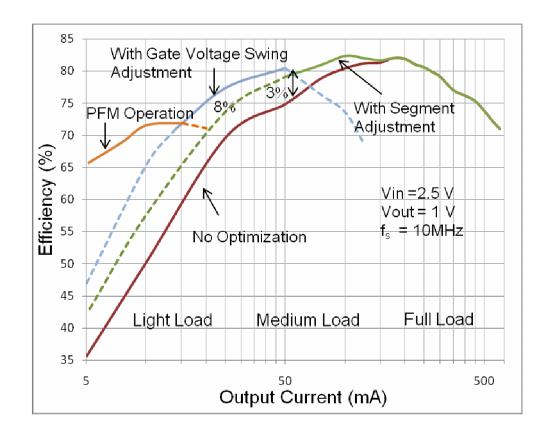
### **IC-Implementation**



Specifications	Value	Units
CMOS Process	0.13	μm
Area	2.5	mm2
Input Voltage	2.5	v
Output Voltage	0.8-1.3	v
Rated Load	500	mA
Filter L,C	400, 0.9	nH,µF
Switching Frequency ,	10	MHz
Ron Pmos , Nmos	0.26 , 0.234	Ω
Supply Analog, Digital	1.2, 2.5	V
Peak Efficiency	83	%
CPM Controller Current	500	μΑ
PFM Controller Current	10	μΑ
Digital Core	200	μΑ

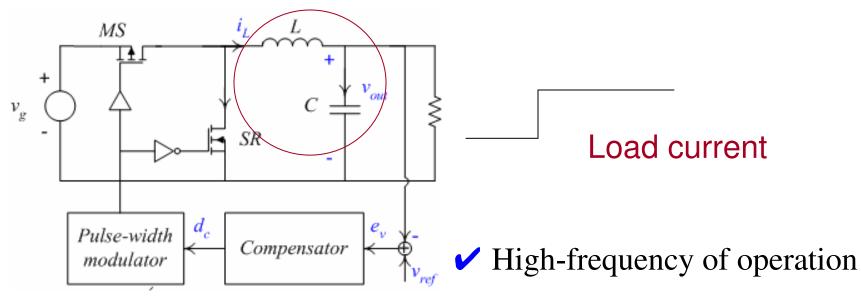


### **Experimental Results**





## Reaching the Physical Limitations Through Hardware-Efficient Mixed-Signal ICs



✓ Optimal response with plug and play operation

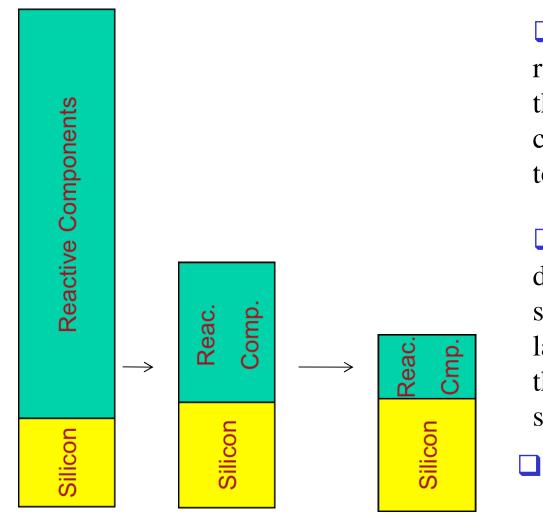
 Relatively flat efficiency curve even during load changes



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### Moving Forward Advanced Low-Volume SMPS Topologies





#### **Moving Forward**

Reduce volume of reactive components through advanced converter and control topologies

Allow weight (cost)
distribution where the
silicon are will be
larger than today but
the overall volume
smaller

□ No penalty in efficiency



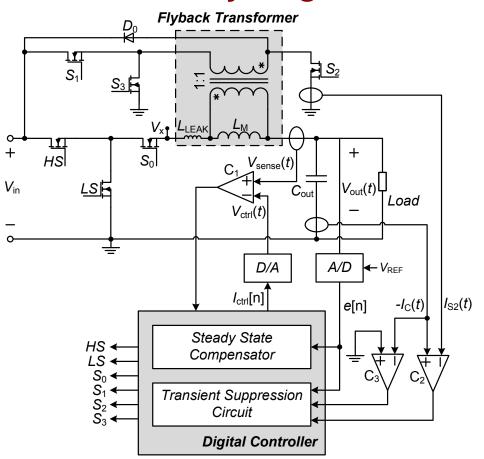
Overall volume (weight) contribution

# Flyback Transformer Based Buck (FTBB) with Transient Energy Recycling<sup>[4]</sup>



[4] Jing Wang; Prodić, A.; Wai Tung Ng, "Mixed-Signal-Controlled Flyback-Transformer-Based Buck Converter With Improved Dynamic Performance and Transient Energy Recycling," IEEE Transactions on Power Electronics, Vol. 28, Issue. 3, February 2013.

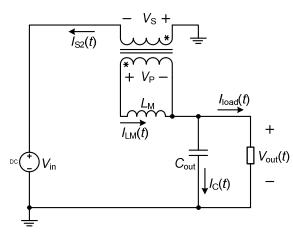
### Flyback Transformer Based Buck with Transient Energy Recycling

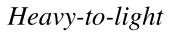




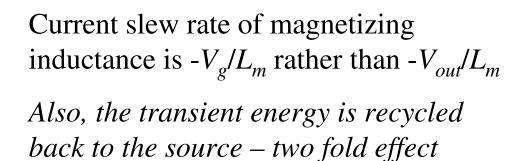
The buck inductor is replaced with a flyback transformer, and a single extra switch inside the conduction path

### **FTBB** – **Principle of Operation**





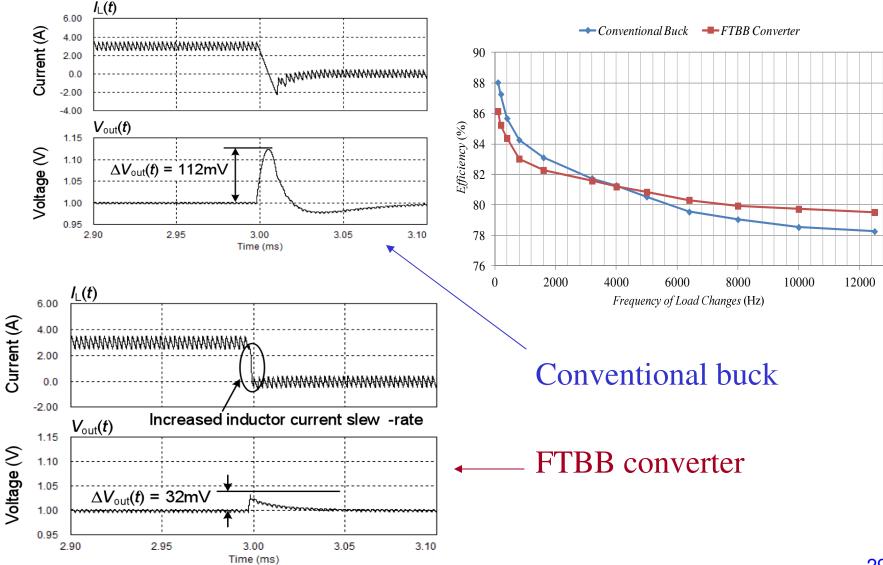
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 $L_{LEAK}$   $I_{load}(t)$  +  $V_{in}$   $V_{in}$   $V_{irtual short}$   $I_{load}(t)$  +  $V_{out}(t)$  - Light-to-heavy

The inductance is reduced to its leakage value

#### **Conventional Minimum-Deviation vs. FTBB**

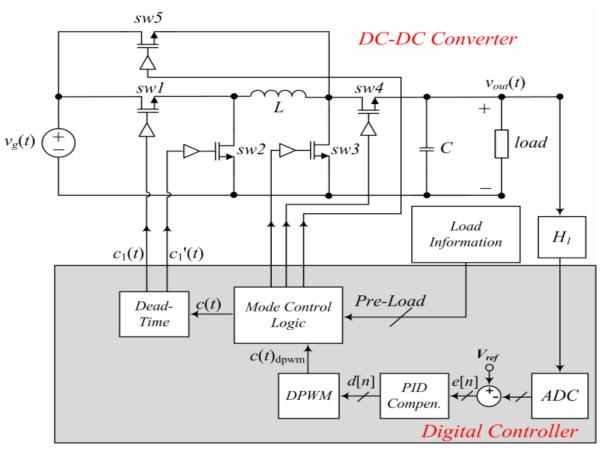


### Load-Interactive Steered-Inductor DC-DC Converter with Transient Energy Recycling <sup>[5]</sup>



[5] S.M. Ahsanuzzaman, A. Parayandeh, A. Prodic, D. Maksimovic, "Load-interactive steered-inductor dc-dc converter with minimized output filter capacitance," in Proc. IEEE Applied Power Electronics Conference (APEC '10), 2010, Pg. 980-985.

### Load Interactive SMPS with Current Steering

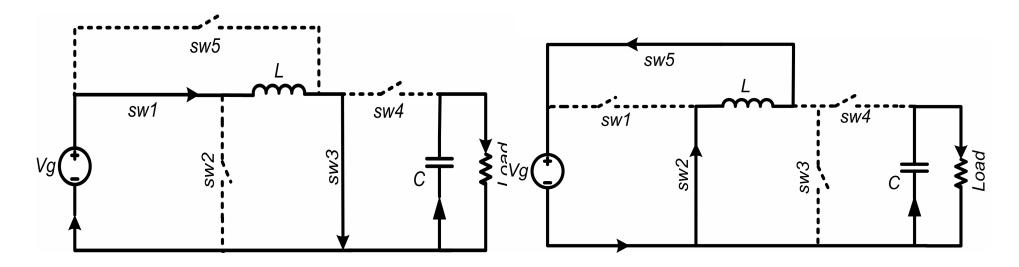




Relies on improved interaction with the digital load

• Theoretically, allows selection of the output capacitor based on the output ripple criteria only.

### **Steered Inductor Buck-Boost Converter**



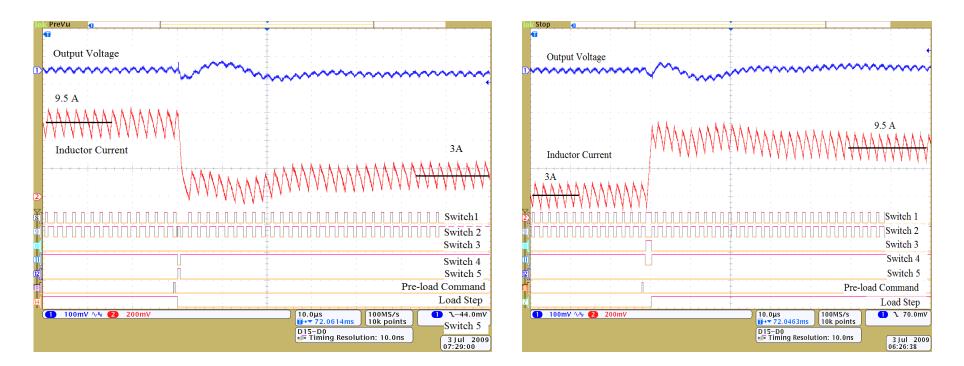
 Light-to heavy pre-transient condition current "pump-up"

Slew rate is Vg/L vs (Vg-Vout)/L

•Heavy-to-light transient ,the current is steered away from the capacitor to the source)

- •The current slew-rate is –Vg/L (was -Vout/L in buck mode)
- Energy recycled in this cycle





#### **Load Transient Performance**



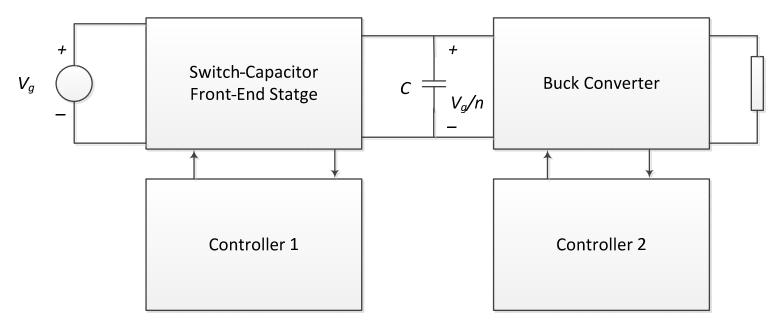
Theoretically, allows reduction of the output capacitor to ripplelimited value

### **Buck Converter with Merged Capacitive Attenuator**<sup>[6]</sup>



[6] A. Radic, A. Prodic, "Buck Converter With Merged Active Charge-Controlled Capacitive Attenuation," IEEE Transactions on Power Electronics, Vol. 27, Issue. 3, March 2012.

### Common Approach: Serial Connection of a Switch Cap Converter And a Converter with Inductor



1. Reduced output filter volume

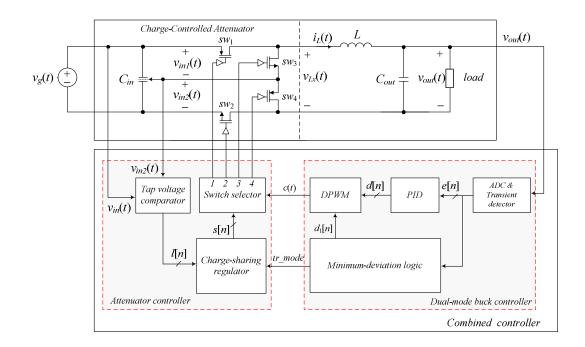
1. Bulky intermediate balancing cap

2. Extra switches in conduction path (at least 4) and at least 6 switches total

3. Requires two control loops



#### **Buck Converter with Merged Capacitive Attenuator**



-All switches rated at  $\frac{1}{2} V_{max}$ of the conventional buck (no extra conduction losses)

-Switches are shared between the cap stage and buck

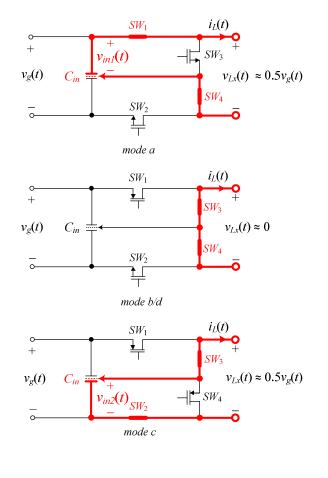
-Lower switching losses than of the conventional buck

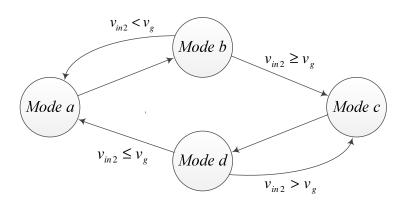
-Centre tap voltage maintained constant with the help of buck inductor

- Better transient response than the time-optimal buck



#### Modes of Operation, Ideal and Practical System

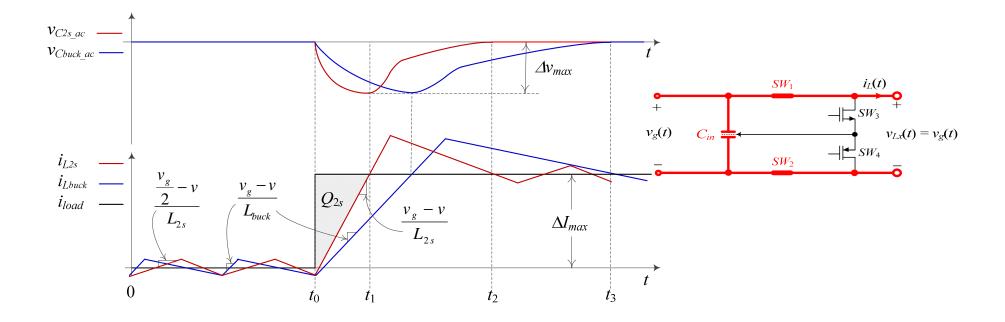




Centre-tap voltage controller operation. Skips regular sequence and takes the charge from the cap with larger voltage until balance is achieved.



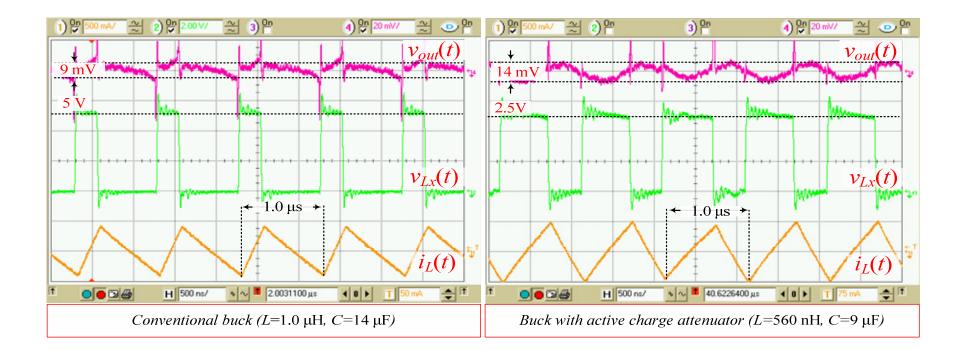
### **Transient Mode**





Comparison with a conventional buck

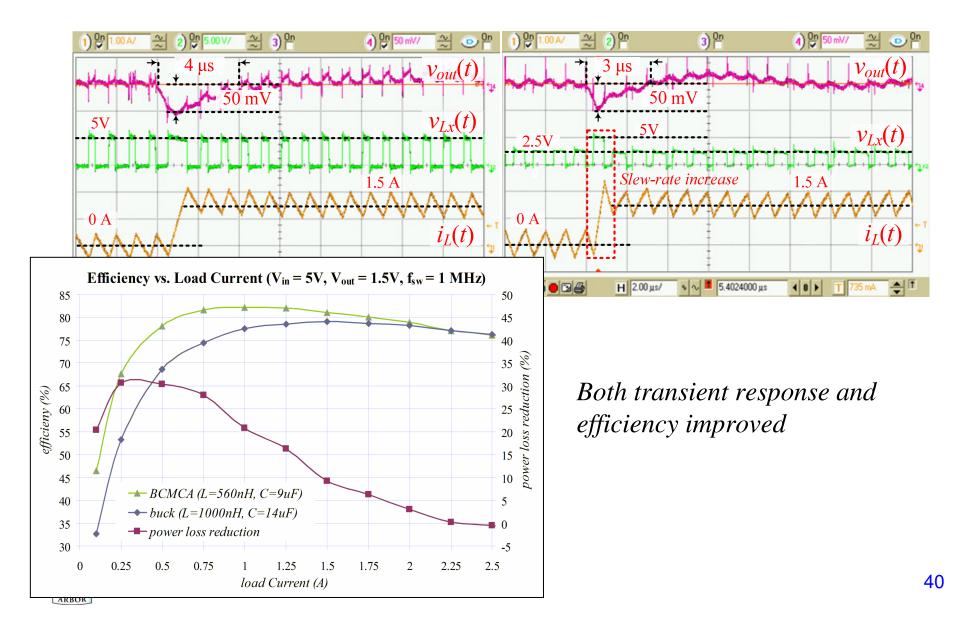
### **Experimental Results: Comparison with Conv. Buck**





For a 5V to 1 V buck 44% smaller inductor and 35% smaller output capacitor

#### **Experimental Results: Comparison with Con. Buck**



#### Extension to 2-Phase<sup>[7]</sup> *2-phase interleaved buck* Input filter converter Active capacitive divider $L_{f}$ $L_p$ $v_{xl}(t)$ $Cr_{-}$ L1G1+ Cin1 $Q_5$ $V_{in1}$ Ē $v_{out}(t)$ G8G9Cout $V_{batt}$ $\mathbb{P}_{\mathbb{P}}Q_3$ *06* ™ $\leq R$ $v_{x2}(t)$ L2+ Cin2 $V_{in2}$ G2G3 G5 G7G4*G6* ADC & Switching control $\leftarrow DPWM \leftarrow d[n]$ e[n]PID . Transient logic detector $d_I[n$ Minimum Deviation. logic Digital controller



 $L_1$  takes the charge from the top inductor and  $L_2$  from the bottom

[7] B. Mahdavikhah, P. Jain, A. Prodic, "Digitally controlled multi-phase buck-converter with merged capacitive attenuator," Applied Power Electronics Conference and Exposition (APEC), 2012 Twenty-Seventh Annual IEEE, vol., no., pp.1083-1087, 5-9 Feb. 2012

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### Many other examples .....



### Conclusion

Digital control allows us to use advanced converter topologies and drastically reduce the volume of SMPS while improving efficiency at the same time.



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### Thank you

